Ordering number:ENN573F

Monolithic Linear IC

LA3161

2-Channel Preamplifier For Car Stereo

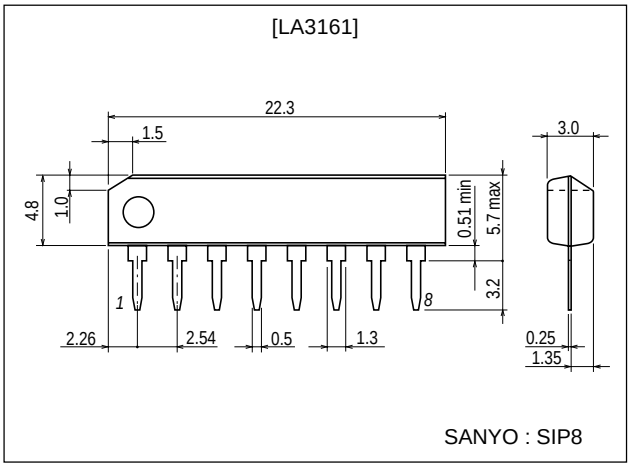
Features

- On-chip 2 preamplifiers.
- Good ripple rejection owing to on-chip voltage regulator.
- Minimum number of external parts required.
- Low noise.
- 8-pin SIP package facilitating easy mounting.
- Pin-compatible with LA3160.

Package Dimensions

unit:mm

3016B-SIP8



Specifications

Absolute Maximum Ratings at Ta = 25˚C

Parameter	Symbol	Conditions	Ratings	Unit
Maximum Supply Voltage	V_{CC} max		18	V
Allowable Power Dissipation	Pd max		200	mW
Operating Temperature	Topr		–20 to +75	˚C
Storage Temperature	Tstg		–40 to +125	˚C

Recommended Operating Conditions at Ta = 25˚C

Parameter	Symbol	Conditions	Ratings	Unit
Supply Voltage	V_{CC}		9	V
Load Resistance	R_L		10k	Ω

Operating Conditions at Ta = 25˚C, V_{CC}=9V, R_L=10kΩ, Rg=600Ω, f=1kHz, NAB

Parameter	Symbol	Conditions	Ratings			Unit
			min	typ	max	
Current Dissipation	I_{CC}			6.5	8.0	mA
Voltage Gain	VG	Closed loop		35		dB
		Open loop, V_O=0.77V	70	78		dB
Output Voltage	V_O	THD=1%	1.0	1.3		V
Total Harmonic Distortion	THD	V_O=0.5V		0.05	0.30	%
Input Resistance	r_i		70k	100k		Ω

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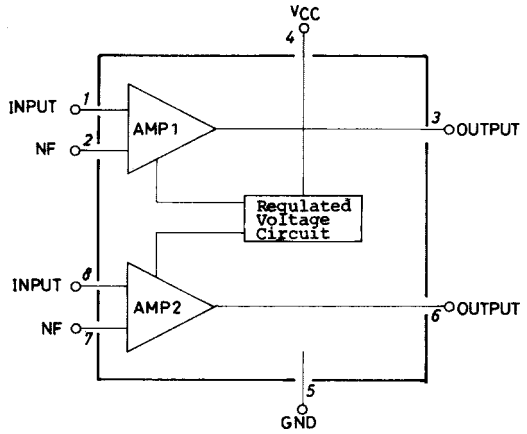
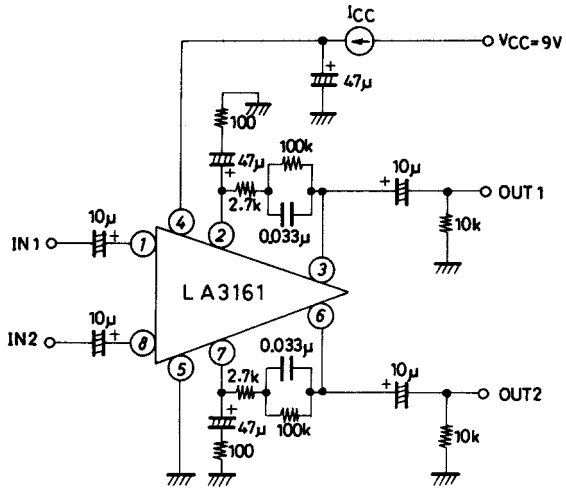
Parameter	Symbol	Conditions	Ratings			Unit
			min	typ	max	
Equivalent Input Noise Voltage	V_{NI}	Rg=2.2kΩ		1.2	2.0	μF
Crosstalk	CT	Rg=2.2kΩ	–50	–65		dB
Ripple Rejection	Rr			–40		dB

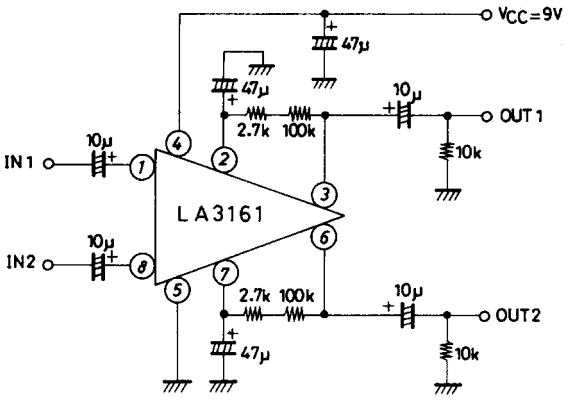
Equivalent Circuit Block Diagram



Test Circuit1 : V_O, VG, THD, I_{CC}, r_i

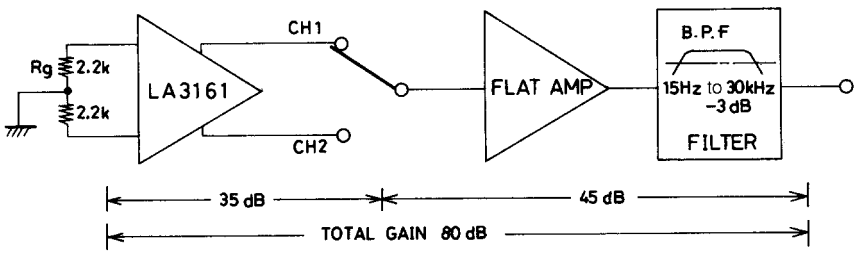


Test Circuit2 : VG_O

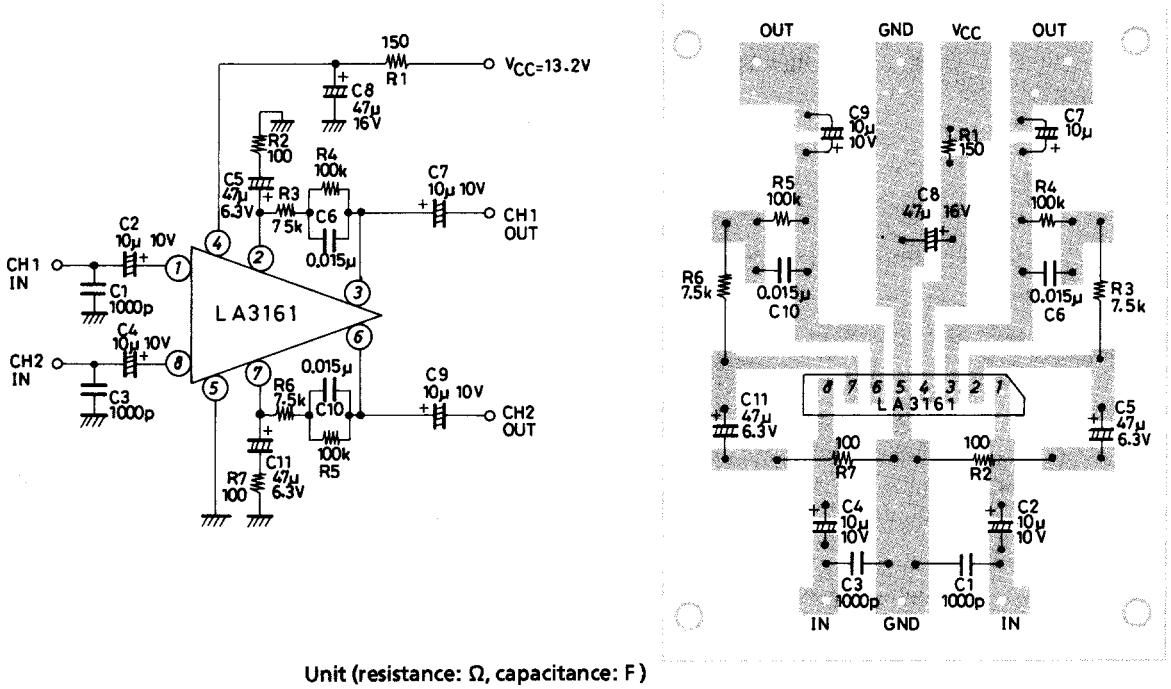


Unit (resistance: Ω, capacitance: F)

Test Circuit3 : Noise



Sample Application Circuit 1 : Preamplifier for Car Stereo



Unit (resistance: Ω, capacitance: F)

Function of External Parts

C_2, C_4 are input coupling capacitors. In NAB equalizer amplifier, the gain at low frequencies is high and 1/f noise inside the IC is emphasized as output noise. Therefore, if the reactance of capacitor at low frequencies is increased, the dependence of 1/f noise on the signal source resistance causes the output noise voltage to deteriorate, and the value of reactance must be made small enough as compared with the signal source resistance. C_2, C_4 also influence the operation start time and the adequate value of these capacitors is 10µF. (Since C_2, C_4 of less than 4.7µF make the operation start time longer, use C_2, C_4, of 4.7µF or more).

C_5, C_{11} are NF capacitors. The lower cut-off frequency depends on the value of these capacitors.

If the lower cut-off frequency is taken as f_L :

$$C5\ (C11) = 1/2\pi \cdot f_L \cdot R2\ (R7)$$

If the value of this capacitor is made larger, the operation start time of amplifier is more delayed. The adequate value of capacitor is 47µF.

The frequency characteristic of the equalizer amplifier depends on C_6 and R_4, R_3 (C_{10} and R_5, R_6).

The time constants to obtain the standard NAB characteristic are as shown below.

Tape speed	9.5cm/s	4.75cm/s
C6 (R3 + R4)	3180µs	1590µs
R3 C6	90µs	120µs

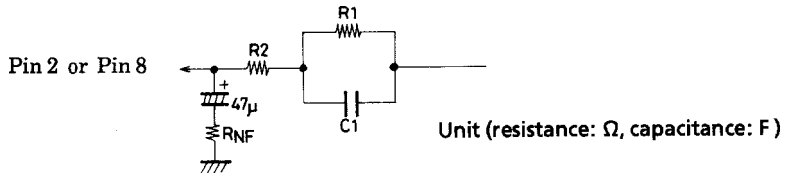
C_8 is bias capacitor for the power line. C_8 of 47µF is inserted at a point as close to the power supply pin (pin 4) as possible.

C_1, C_3 are for preventing radio interference in the strong electric field, interference attributable to engine noise, and blocking oscillation at the time of large amplitude operation. The adequate value of C_1, C_3 is approximately 1000pF.

C_7, C_9 are output coupling capacitors. The adequate value of C_7, C_9 is 10µF.

NAB element and determination of gain

Since the DC feedback is provided by R_1, R_2 of NAB element, which brings about DC output potential at pins 3, 6, it is impossible to change the value of R_1, R_2 of NAB element greatly. Therefore, when determining the gain, change R_{NF} with R_1, R_2, C_1 (NAB element) kept constant.



(1) How to obtain R_{NF}

Impedance Z of NAB element is

$$Z = \frac{1}{1/R1+j\omega C1} + R2$$

$$= (R1+R2)\left\{\frac{1+j\omega C1\ \{R1R2/\ (R1+R2)\}}{1+j\omega C1R1}\right\}$$

For a general negative feedback amplifier circuit, $A=Ao/(1+Ao\beta)$ applies, and $Z=A \cdot R_{NF}$ is obtained under conditions of $Ao>>A$, $A>>1$ ($\beta=R_{NF}/(R_{NF}+Z)$, Ao=open-loop gain, A=feedback gain).
Therefore, we can use an approximation of $R_{NF}=Z/A$.

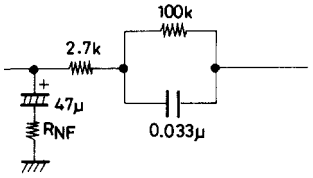
A= (VG for 1kHz) times, (Set R_1, R_2 at approximately 100kΩ)

Each time constant of NAB characteristic.

Tape speed	9.5cm/s	4.75cm/s
T1 C1, R1	3180µs	1590µs
T2 C1 (R1//R2)	90µs	120µs

(2) Examples of NAB Constants

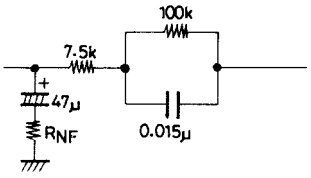
(a) Tape speed : 9.5cm/s. (8 tracks)



VG : R_{NF} (VG/f=1kHz)

VG	30	35	40	dB
R_{NF}	180	100	56	Ω

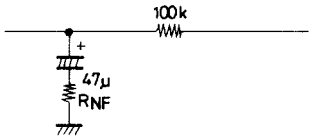
(b) Tape speed : 4.75cm/s. (cassette)



VG : R_{NF}

VG	30	35	40	dB
R_{NF}	440	240	130	Ω

(c) Flat amplifier



VG : R_{NF}

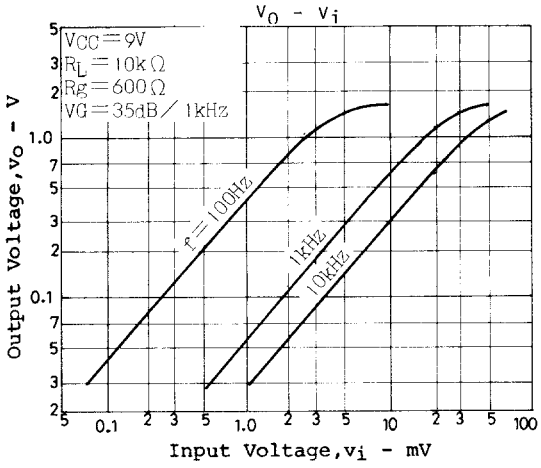
VG	30	35	40	dB
R_{NF}	3.2	1.8	1	kΩ

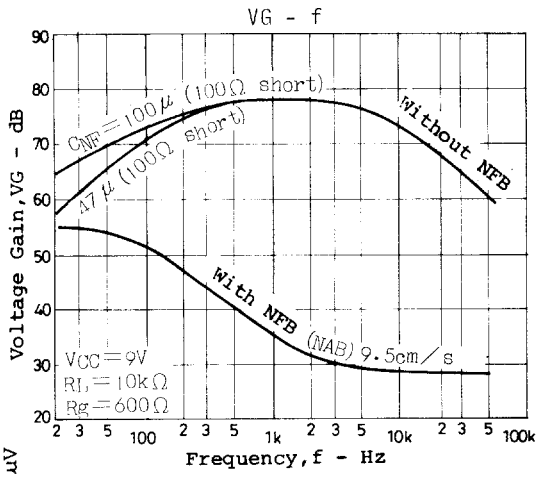
Proper cares in using IC

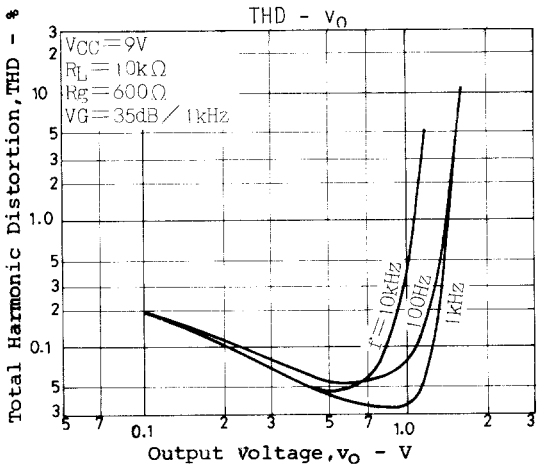
1. Maximum Rating
   If the IC is used in the vicinity of the maximum rating, even a slight variation in conditions may cause the maximum rating to be exceeded, thereby leading to a breakdown. Allow an ample margin of variation for supply voltage, etc. and use the IC in the range where the maximum rating is not exceed.
2. Short between pins
   If the supply voltage is applied when the space between pins is shorted, a breakdown or deterioration may occur. When installing the IC on the board or applying the supply voltage, make sure that the space between pins is not shorted with solder, etc.
3. Breakdown of IC attributable to inverted insertion
   If the IC is inserted inversely and operated, the IC may suffer from something unusual, thereby leading to a breakdown or deterioration of the IC. When installing the IC on the board or operating the IC, check the marked surface of IC.

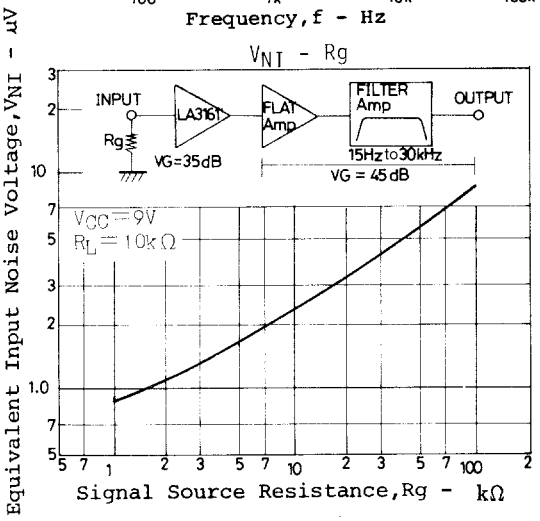
Proper cares to be taken for obtaining optimum operation of IC

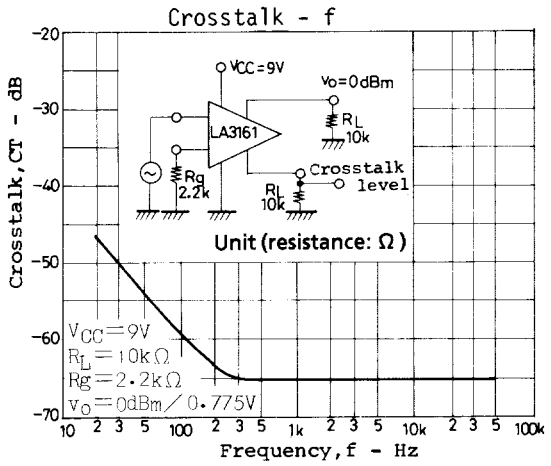
· Set DC resistance of R_1, R_2 of NAB element at approximately 100kΩ.
· Determine the gain by changing R_{NF} without chaging NAB constant (Refer to Examples of NAB constant.).

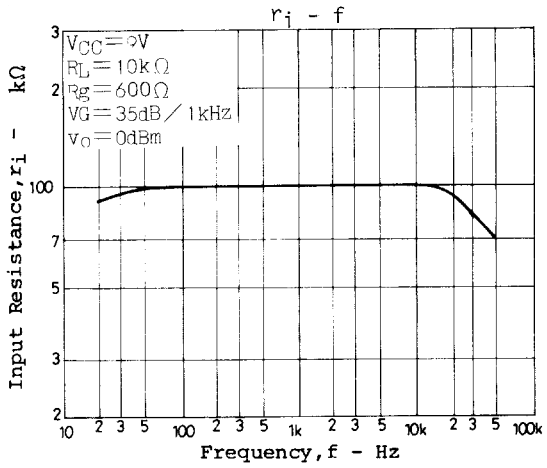


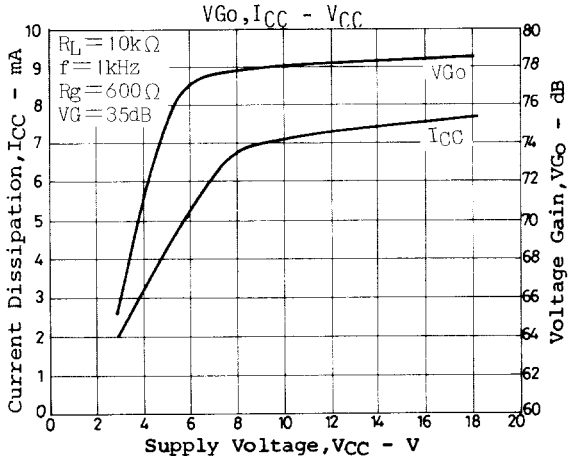


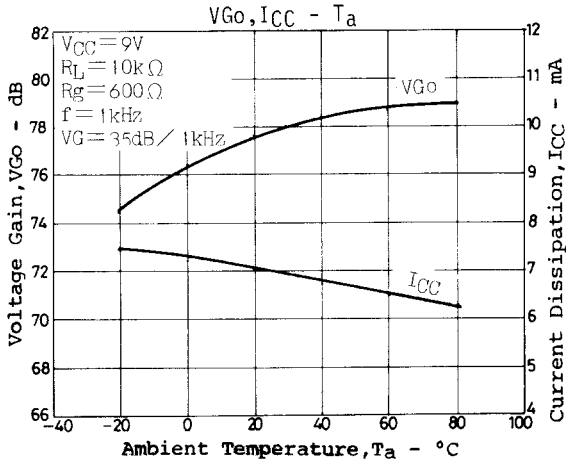


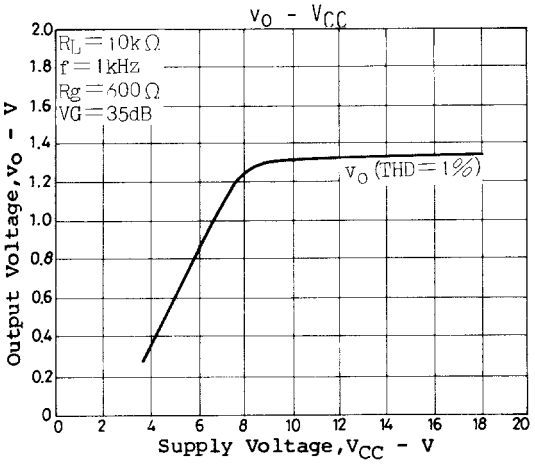


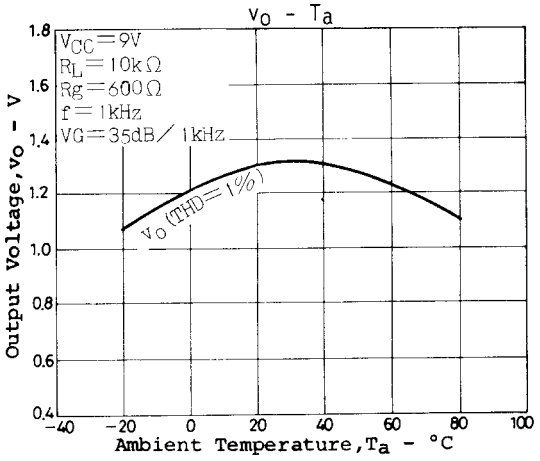

Crosstalk - f
VCC=9V
Vo=0dBm
RL 10k
LA3161
Rg 2.2k
RL 10k
Crosstalk level
Unit (resistance: Ω)
VCC=9V
RL=10kΩ
Rg=2.2kΩ
Vo=0dBm/0.775V
Crosstalk,CT - dB
Frequency,f - Hz

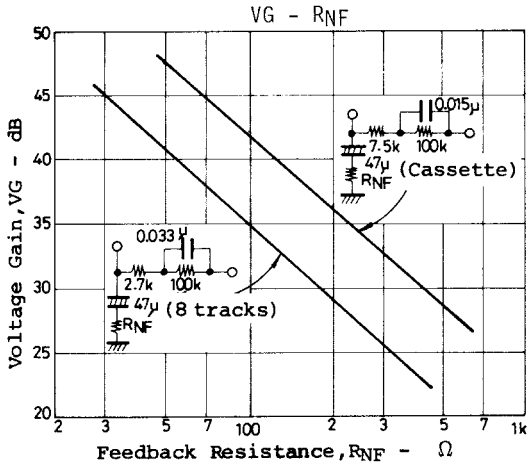

ri - f
VCC=9V
RL=10kΩ
Rg=600Ω
VG=35dB/1kHz
Vo=0dBm
Input Resistance,ri - kΩ
Frequency,f - Hz

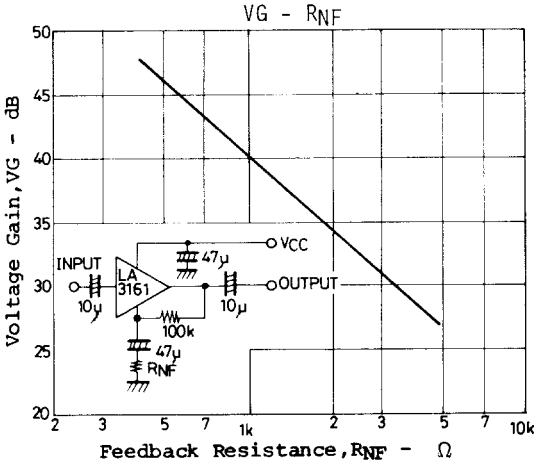

VGo,ICC - VCC
RL=10kΩ
f=1kHz
Rg=600Ω
VG=35dB
VGo
ICC
Current Dissipation,ICC - mA
Voltage Gain,VGo - dB
Supply Voltage,VCC - V

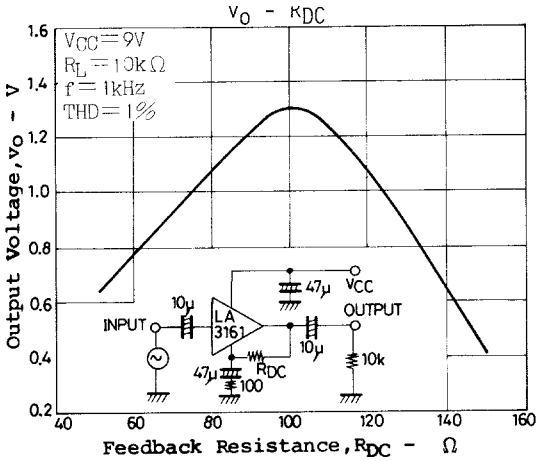

VGo,ICC - Ta
VCC=9V
RL=10kΩ
Rg=600Ω
f=1kHz
VG=35dB/1kHz
VGo
ICC
Voltage Gain,VGo - dB
Current Dissipation,ICC - mA
Ambient Temperature,Ta - °C


Vo - VCC
RL=10kΩ
f=1kHz
Rg=600Ω
VG=35dB
Vo (THD=1%)
Output Voltage,Vo - V
Supply Voltage,VCC - V


Vo - Ta
VCC=9V
RL=10kΩ
Rg=600Ω
f=1kHz
VG=35dB/1kHz
Vo (THD=1%)
Output Voltage,Vo - V
Ambient Temperature,Ta - °C


VG - RNF
0.015μ
7.5k
100k
47μ (Cassette)
RNF
0.033μ
2.7k
100k
47μ (8 tracks)
RNF
Voltage Gain,VG - dB
Feedback Resistance,RNF - Ω


VG - RNF
INPUT
10μ
LA 3161
47μ
VCC
OUTPUT
10μ
100k
47μ
RNF
Voltage Gain,VG - dB
Feedback Resistance,RNF - Ω



Unit (resistance: Ω, capacitance: F)